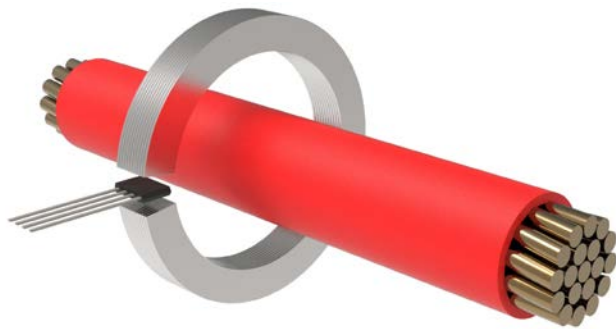


1 Product Description

The MagnTek® MT9511 product series is a monolithic programmable Hall effect linear sensor IC. The device can be used for accurate position sensing in a wide range of applications.

Each of the MT9511 consists of a highly sensitive Hall element, a low noise small-signal high-gain amplifier, a clamp and overcurrent protection output stage, and a high bandwidth dynamic offset cancellation technique.

The MT9511 provides an analog output voltage proportional to the applied magnetic flux density. The customer can configure the sensitivity and quiescent (zero field) output voltage through programming on the output pins, to optimize performance in the end application. The quiescent output voltage is user-adjustable around 50% of the supply voltage(VCC) and the output sensitivity is adjustable within the range of 0.6 to 22.4mV/G.



2 Features

- End-of-line programmable
- Typical Accuracy: --- $\pm 1.0\%$ (25°C)
- High Linearity: --- $\pm 0.2\%$ (25°C)
- High Bandwidth: --- 250kHz
- Wide Operating Temperature: --- -40°C~150°C
- Fast Output Response Time: --- 2.2 μs (typ.)
- Package Option: ---SIP-4
- High stability over operation temperature range: --- $\pm 2.0\%$ (25°C~150°C)
--- $\pm 2.5\%$ (-40°C~25°C)
- Ratiometric Output from Supply Voltage
- Low-Noise Analog Signal Path
- RoHS Compliant: (EU)2015/863

3 Applications

- Inverter current sensing
- Motor phase and rail current sensing
- PV string inverters
- Battery management system
- Switching power supplies
- Overcurrent protection

4 Product Overview of MT9511A

Part Number	Sensitivity Range	Package	Packing
MT9511A-01	0.6~1.4 mV/Gs	SIP-4	Bulk packaging(500pcs/bag)
MT9511A-02	1.4~2.8 mV/Gs	SIP-4	Bulk packaging(500pcs/bag)
MT9511A-04	2.8~5.6 mV/Gs	SIP-4	Bulk packaging(500pcs/bag)
MT9511A-08	5.6~11.2 mV/Gs	SIP-4	Bulk packaging(500pcs/bag)
MT9511A-16	11.2~22.4 mV/Gs	SIP-4	Bulk packaging(500pcs/bag)

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Reversion History

1	Originally Version	
2	1.1 Version	Update the definition of TPO; add POR parameters
3	1.2 Version	Corrected the marking error of package information
4	1.3 Version	Update Bulk packaging

5 Functional Block Diagram

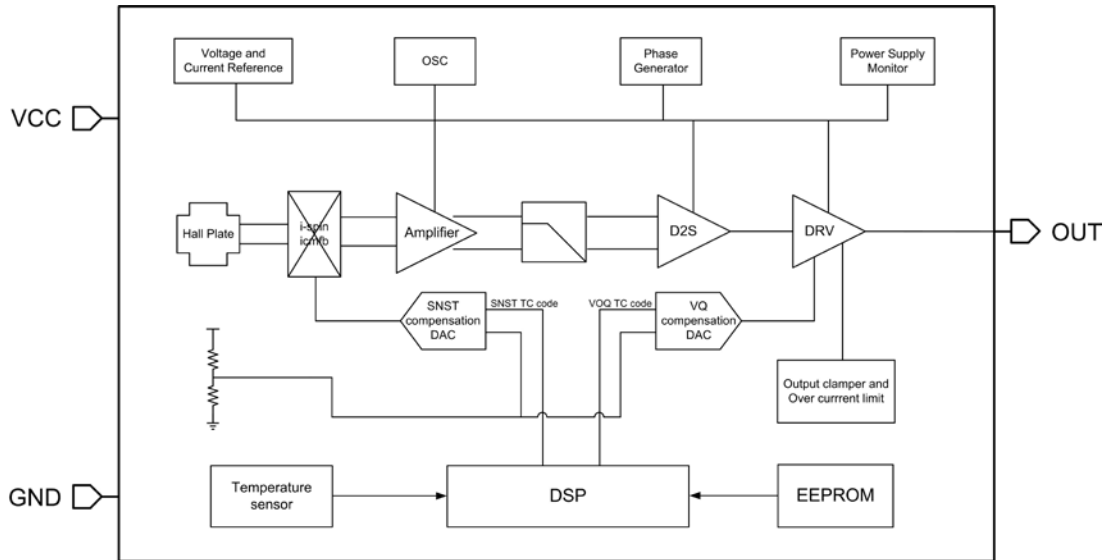


Figure.1 Functional Block Diagram

6 Pin Configuration and Functions

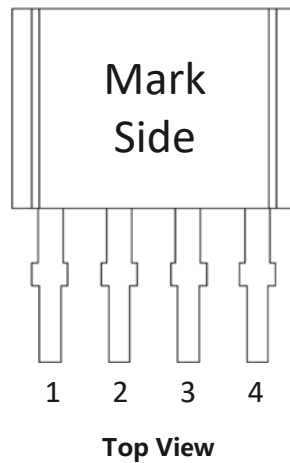


Figure. 2 Pin Configuration & Functions

No.	Name	Description
1	VCC	Power Supply
2	VOUT	Analog Output Signal
3	NC	No Connect
4	GND	Signal Ground

7 Transfer Characteristics

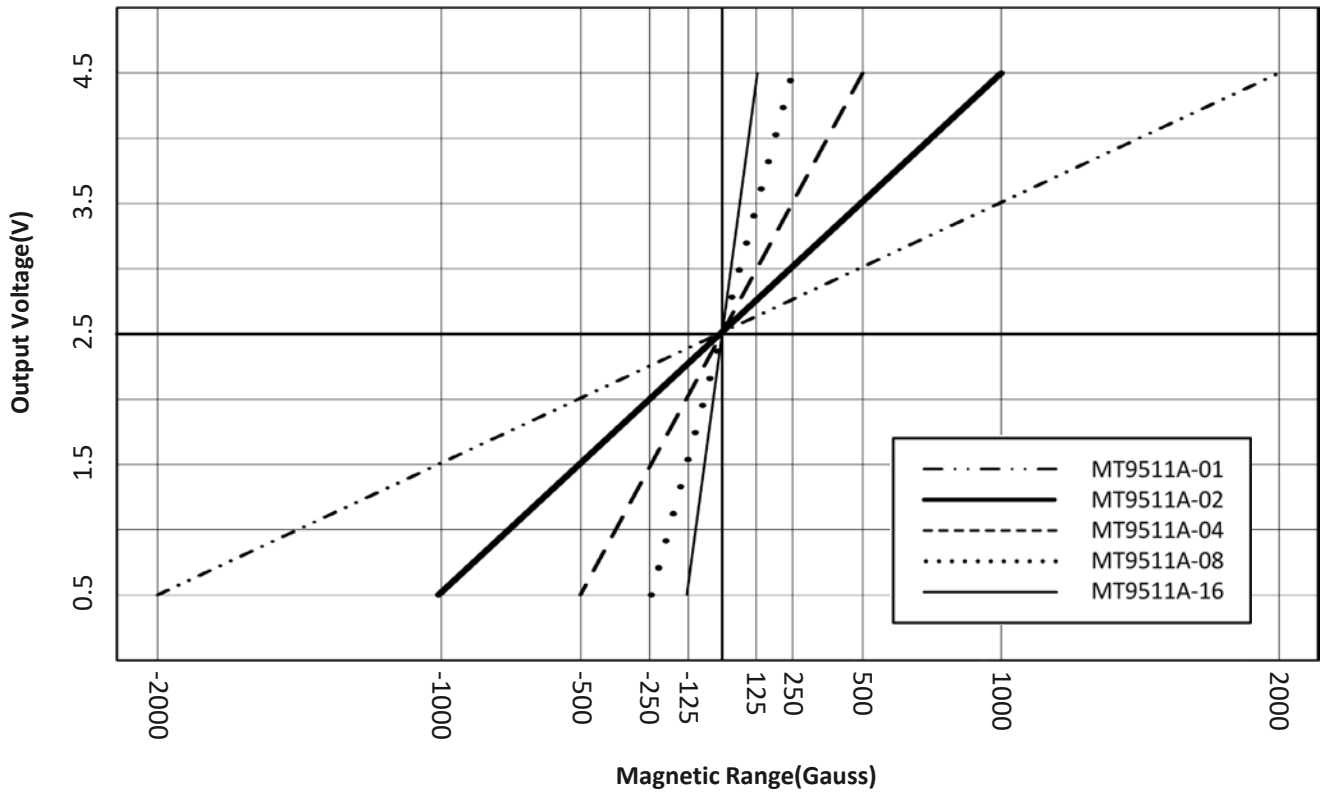


Figure.3 Transfer Characteristics

8 Typical Application Circuit

The typical application circuits of MT9511series products include a bypass capacitor and a filter capacitor as an additional external components. **CBYPASS capacitor between VCC and GND is necessary.** Magnetic field applied vertically to chip surface, the analog signal output is measured directly from the VOUT pin

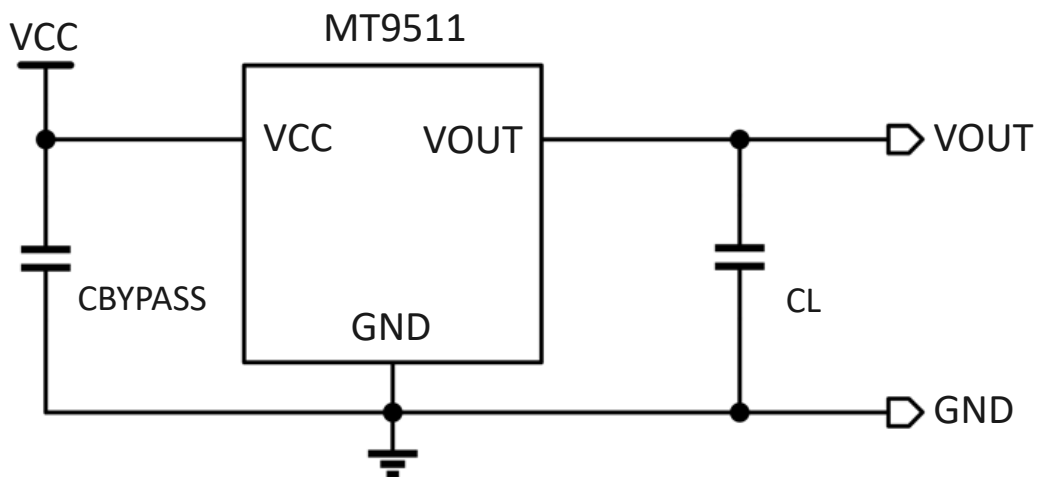


Figure.4 Typical Application Circuit

9 Electrical Magnetic Characteristics

9.1 Absolute Maximum Ratings

Absolute maximum ratings are limited values to be applied individually, and beyond which the serviceability of the circuit may be impaired. Functional operability is not necessarily implied. Exposure to absolute maximum rating conditions for an extended period of time may affect device reliability.

Symbol	Parameters	Min	Max	Units
VCC	Supply Voltage	-	6	V
VRCC	Reverse Battery Voltage	-0.1	-	V
VOUT	Output Voltage	-	VCC+0.5	V
VROUT	Reverse Output Voltage	-0.1	-	V
IOUT(source)	Continuous Output Current(source)	-	55	mA
IOUT(sink)	Continuous Output Current(sink)	-	55	mA
TA	Operating Ambient Temperature	-40	150	°C
TS	Storage Temperature	-50	150	°C
TJ	Junction Temperature	-	165	°C
Endurance	Number of EEPROM Programming Cycles	200	-	cycle

9.2 ESD Ratings

Symbol	Parameters	Reference	Values
VESD	Human-body model(HBM)	AEC-Q100-002	Class IIIA
	Charged-device model(CDM)	AEC-Q100-011	Class C6
	Latch up (Latch up)	AEC-Q100-004	Class IIA

9.3 Electrical Specifications

TA=-40~150 °C, VCC=5V, CBYPASS=0.1uF (unless otherwise specified)

Symbol	Parameters	Test Condition	Min	Typ	Max	Unit
VCC	Supply Voltage	-	4.5	5	5.5	V
ICC	Supply Current	TA = 25°C	-	12	18	mA
BW	Internal Bandwidth	-3 dB; CL = 1 nF	-	250	-	KHz
TPO	Power on time	TA = 25°C, no CBYPASS, CL = 1nF	-	1.3	-	ms
VUVLOH	Undervoltage Lockout(UVLO) High Voltage	TA = 25°C, VCC rising and device function enabled	-	4	-	V
VUVLOL	Undervoltage Lockout(UVLO) Low Voltage	TA = 25°C, VCC falling and device function disabled	-	3.75	-	V
VUVLOHYS	UVLO Hysteresis	TA = 25°C	-	0.25	-	V
TUVLOD	UVLO Delay Time	TA = 25°C	-	30	-	us
VPORH	Power-On Reset High Voltage	TA = 25°C, VCC rising	-	2.75	-	V
VPORL	Power-On Reset Low Voltage	TA = 25°C, VCC falling	-	2.55	-	V
VPORHYS	Power-On Reset Hysteresis	TA = 25°C	-	0.2	-	V
ISCLP	Source Current of Over-Current-Limit	-	-	55	-	mA
ISCLN	Sink Current of Over-Current-Limit	-	-	55	-	mA
TSCLD	Detect Time for over-Current-Limit	TA = 25°C, IOOUT>ISCLP or IOOUT<ISCLN	-	10	-	us
TSCLR	Release Time for over-Current-Limit	TA = 25°C	-	1	-	ms
VOL	Analog Output Low Saturation Level	RL>=4.7KΩ	-	-	0.3	V
VOH	Analog Output High Saturation Level	RL>=4.7KΩ	VCC-0.3	-	-	V
CL	Output Cap Load	OUT to GND	-	0.47	1	nF
RL	Output Res Load	Pull-down to GND	4.7	-	-	KΩ
		Pull-up to VCC	4.7	-	-	KΩ
ROUT	DC Output resistance	TA=25°C	-	5	-	Ω
TR	Rise time	B = B(max), TA = 25°C, CL = 1nF	-	1.8	-	us
TPD	Propagation Delay	B = B(max), TA = 25°C, CL = 1nF	-	1.4	-	us
TRESP	Response Time	B = B(max), TA = 25°C, CL = 1nF	-	2.2	-	us

Continued on the next page...

Electrical Specifications(continued)T_A=-40~150 °C, V_{CC}=5V, C_{BYPASS}=0.1uF (unless otherwise specified)

Symbol	Parameters	Test Condition	Min	Typ	Max	Unit
VCLP_LO	Clamp Low Output Level	TA = 25°C, RL = 10kΩ to VCC	0.15	-	0.45	V
VCLP_HI	Clamp High Output Level	TA = 25°C, RL = 10kΩ to GND	4.55	-	4.85	V
TCLP	Clamp Low Output Level	TA=25°C, magnetic field step from: 800 to 1200Gs, CL=1nF, SNST=2 mV/Gs	-	8	-	us
IND	Noise Density	Input-referenced noise density; TA = 25°C, SNST=6.88mV/Gs	-	1	-	mG/√Hz

Accuracy Specification

ELIN	Nonlinearity Sensitivity Error	TA = 25°C, VCC=5V	-0.5	±0.2	0.5	%
VOQ	Quiescent Voltage Output Error	TA = 25°C, VCC=5V	-10	-	10	mV
SNST_INIT	Initial Unprogrammed Sensitivity	TA = 25°C, VCC=5V	0.98	1	1.02	mV/Gs
			0.96	2	2.04	mV/Gs
			3.92	4	4.08	mV/Gs
			7.84	8	8.16	mV/Gs
			15.68	16	16.32	mV/Gs
ERAT_SNST	Ratiometry Sensitivity Error	VCC = 4.5 ~ 5.5 V, TA = 25°C	-	±1.5	-	%
ERAT_VOQ	Ratiometry Quiescent Voltage Output Error	VCC = 4.5 ~ 5.5 V, TA = 25°C	-	±1	-	%
ERAT_CLP	Ratiometry Clamp Error	VCC = 4.5 ~ 5.5 V, TA = 25°C	-	±1	-	%
ΔSNST_PKG	Sensitivity Drift Due to Package Hysteresis	TA = 25°C, temperature cycling: from 25°C to 150°C and back to 25°C	-	±1.25	-	%

Programming Specification

VOQ_STEP	Average Quiescent Voltage Output Programming Step Size	TA = 25°C, VCC=5V	-	±1.25	-	mV
EVOQ_STEP	Quiescent Voltage Output Programming Resolution	TA = 25°C, VCC=5V	-	±0.625	-	mV
SNST_PR	Sensitivity Programmed Range	TA = 25°C, VCC=5V	0.6	-	1.4	mV/Gs
			1.4	-	2.8	mV/Gs
			2.8	-	5.6	mV/Gs
			5.6	-	11.2	mV/Gs
			11.2	-	22.4	mV/Gs

Continued on the next page ...

Electrical Specifications(continued)

T_A=-40~150 °C, V_{CC}=5V, C_{BYPASS}=0.1uF (unless otherwise specified)

Symbol	Parameters	Test Condition	Min	Typ	Max	Unit
SNST_STEP	Average Sensitivity Programming Step Size	T _A = 25°C, V _{CC} =5V	-	±0.3125	-	%
ESNST_STEP	Sensitivity Programming Resolution	T _A = 25°C, V _{CC} =5V	-	±0.1562	-	%

Factory Temperature Coefficient Programed Specification

ΔSNST_TC	Sensitivity Drift Through Temperature Range	T _A = 25°C to 150°C	-2.0	-	2.0	%
		T _A = -40°C to 25°C	-2.5	-	2.5	%
SNST_TC_STEP	Average Sensitivity Temperature Compensation Step Size		-	±0.07	-	%/°C
ΔVOQ_TC	Quiescent Voltage Output Drift Through Temperature Range	T _A = 25°C to 150°C	-15	-	15	mV
		T _A = -40°C to 25°C	-15	-	15	mV
VOQ_TC_STEP	Average Quiescent Voltage Output Temperature Compensation Step Size		-	1.25	-	mV/°C

Lock Bit Programming

EELOCK_BIT	EEPROM Lock Bit		-	1	-	bit
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10 Characteristic Definitions

Power On Time---TPO

When the supply is ramped to its operating voltage, the device requires a finite time to power its internal components before responding to an input magnetic field.

The Power-On Time (TPO) is defined as the time taken between the supply reaching the minimum operating voltage V_{CCmin} (t_1), and the output voltage to settling to within $\pm 10\%$ of its steady state value under an applied magnetic field (t_2) (See Figure 5).

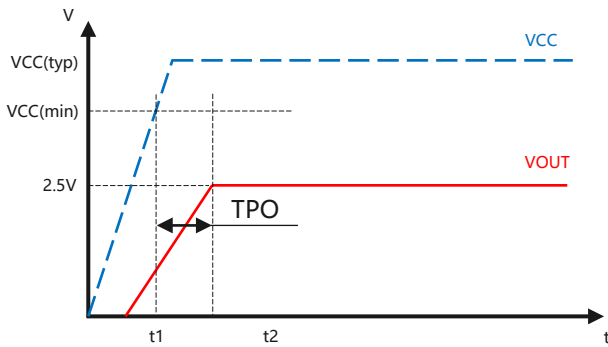


Figure.5 Power On Time Definition

Propagation Delay---TPD

The time interval between a) when the primary current signal reaches 20% of its final value, and b) when the output reaches 20% of its final value (see Figure 6).

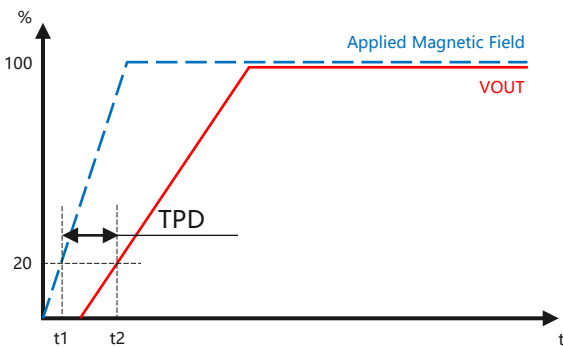


Figure.6 Propagation Delay Definition

Rise Time---TR

Rise Time is the time interval between the sensor VOUT reaching 10% of its full scale value (t_1), and it reaching 90% of its full scale value (t_2). (see Figure 7). Both TR and TRESP can be negatively affected by any eddy current losses created if a conductive ground plane is used.

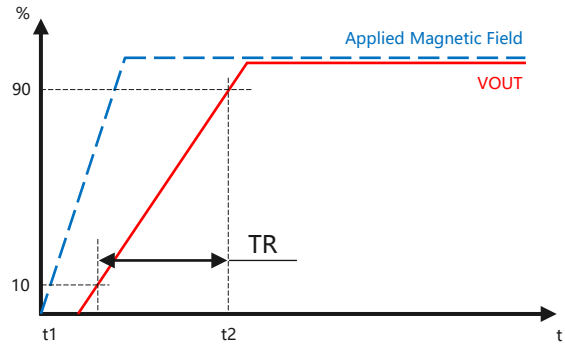


Figure.7 Rise Time Definition

Response Time---TRESP

The time interval between a) when the primary current signal reaches 80% of its final value, and b) when the sensor reaches 80% of its output corresponding to the applied current. (see Figure 8). Both TR and TRESP can be negatively affected by any eddy current losses created if a conductive ground plane is used.

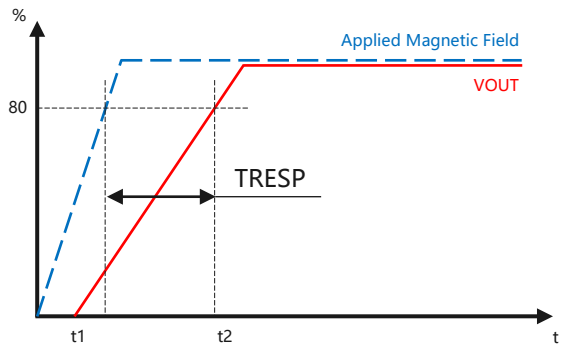


Figure.8 Response Time Definition

Delay to Clamp---TCLP

A large magnetic input step may cause the clamp to overshoot its steady state value. The Delay to Clamp (TCLP) is defined as the time it takes for the output voltage to settle within $\pm 1\%$ of its steady state value, after initially passing through its steady state voltage (see Figure 9).

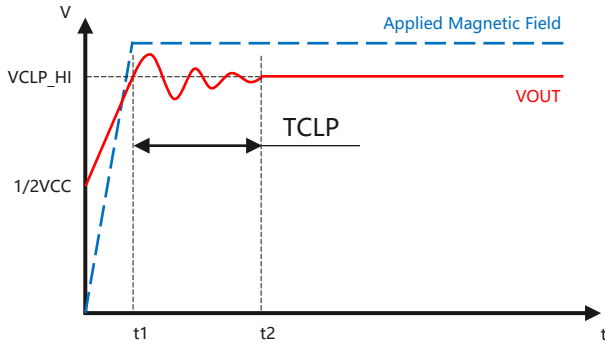


Figure.9 Propagation Delay Definition

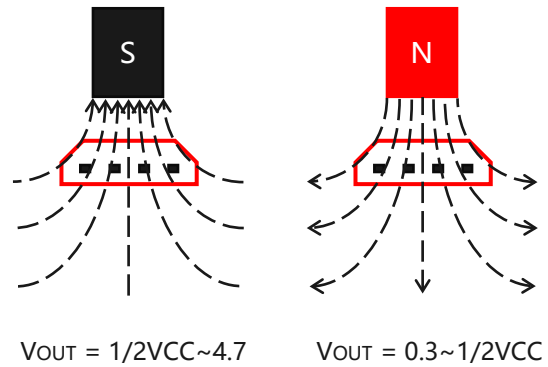


Figure.10 Flux Direction Polarity

Quiescent Voltage Output---VOQ

In the quiescent state (no significant magnetic field: B = 0GS), the output (VOQ), has a constant ratio to the supply voltage (VCC), throughout the entire operating ranges of VCC and ambient temperature (TA).

Quiescent Voltage Output Drift Through Temperature Range---ΔVOQ_TC

Due to internal component tolerances and thermal considerations, the Quiescent Voltage Output (VOQ), may drift from its nominal value through the operating ambient temperature (TA). The Quiescent Voltage Output Drift Through Temperature Range, Δ VOQ_TC, is defined as:

$$\Delta VOQ_{TC} = VOQ(TA) - VOQ_EXPECT(TA)$$

VOQ_TC should be calculated using the actual measured values of VOQ(TA) and VOQ_EXPECT(TA) rather than programming target values

Sensitivity---SNST

The presence of a south polarity magnetic field, perpendicular to the branded surface of the package face, increases the output voltage from its quiescent value toward the supply voltage rail. The amount of the output voltage increase is proportional to the magnitude of the magnetic field applied.

Conversely, the application of a north polarity field decrease the output voltage from its quiescent value. This proportionality is specified as the magnetic sensitivity(Sens(mV/G)),of the device, and it is defined as:

$$SNST = \frac{V_{OUT(BPOS)} - V_{OUT(BNEG)}}{BPOS - BNEG}$$

where BPOS and BNEG are two magnetic fields with opposite polarities.

Sensitivity Drift Through Temperature Range---ΔSNST_TC

Second order sensitivity temperature coefficient effects cause the magnetic sensitivity, to drift from its expected value over the operating ambient temperature range (TA). The Sensitivity Drift Through Temperature Range, Δ SNST_TC, is defined as:

$$\Delta SNST_{TC} = \frac{SNST(TA) - SNST_EXPECT(TA)}{SNST_EXPECT(TA)} * 100\%$$

Sensitivity Drift Due to Package Hysteresis---ΔSNST_PKG

Second order sensitivity temperature coefficient effects cause the magnetic sensitivity, to drift from its expected value over the operating ambient temperature range (TA). The Sensitivity Drift Through Temperature Range(Δ SNST_TC) is defined as:

$$\Delta SNST_{PKG} = \frac{SNST_{25^{\circ}C_2} - SNST_{25^{\circ}C_1}}{SNST_{25^{\circ}C_1}} * 100\%$$

where SNST_25°C_1 is programmed value of sensitivity at TA=25°C, and SNST_25°C_2 is the value of sensitivity at TA=25°C, after temperature cycling from TA to 150°C/168 hours and back to 25°C

Nonlinearity Sensitivity Error---ELIN

Ideally input magnetic field vs sensor output function is a straight line. The non-linearity is an indication of the worst deviation from this straight line. The ELIN in % is defined as:

$$ELIN = \left(\frac{SNST_{B1}}{SNST_{B2}} - 1 \right) * 100\%$$

Where:

$$SNST_B1 = \left(\frac{VOUT_BPOS1 - VOUT_BNEG1}{BPOS1 - BNEG1} \right)$$

$$SNST_B2 = \left(\frac{VOUT_BPOS2 - VOUT_BNEG2}{BPOS2 - BNEG2} \right)$$

and BPOSx and BNEGx are positive and negative magnetic fields, with respect to the quiescent voltage output such that |BPOS2| = |BNEG2| = Bmax, and |BPOS2| = 2 * |BPOS1| and |BNEG2| = 2 * |BNEG1|.

Symmetry Sensitivity Error---ESYM

The magnetic sensitivity of MT9511 device is constant for any applied magnetic fields of equal magnitude and opposite polarities. Symmetry Error (ESYM) is measured and defined as:

$$ESYM = \left(\frac{SNST_BPOSx}{SNST_BNEGx} - 1 \right) * 100\%$$

Where:

$$SNST_BPOSx = \frac{VOUT_Bx - VOQ}{Bx}$$

$$SNST_BNEGx = \frac{VOQ - VOUT_Bx}{Bx}$$

BPOSx and BNEGx are positive and negative magnetic fields such that |BPOSx| = |BNEGx|.

Ratiometry Error---ERAT

The MT9511 device features ratiometric output. This means that the Quiescent Voltage Output (VOQ), magnetic sensitivity (SNST) and Output Voltage Clamp (VCLP_HI, VCLP_LO), are proportional to the Supply Voltage (VCC). In other words, when the VCC increases or decreases by a certain percentage, each characteristic also increases or decreases by the same percentage. Error is the difference between the measured change in the VCC relative to 5V, and the measured change in each Characteristic

Ratiometry Quiescent Voltage Output Error---ERAT_VOQ

ERAT_VOQ, for a given supply voltage, is defined as:

$$ERAT_VOQ = \left(\frac{VOQ(VCC)/VCC}{VOQ(5V)/5V} - 1 \right) * 100\%$$

Ratiometry Sensitivity Error--ERAT_SNST

ERAT_SNST, for a given supply voltage, is defined as:

$$ERAT_SNST = \left(\frac{SNST_B1(VCC)/VCC}{SNST_B1(5V)/5V} - 1 \right) * 100\%$$

Ratiometry Clamp Error---ERAT_CLP

ERAT_CLP, for a given supply voltage, is defined as:

$$ERAT_CLP = \left(\frac{VCLP(VCC)/VCC}{VCLP(5V)/5V} - 1 \right) * 100\%$$

Where VCLP is either VCLP_HI or VCLP_LO.

Over Current Limit---ISCLP & ISCLN

The MT9511 has over-current protection function. When IOU ≥ ISCLP or ISCLN, the output driver will be closed and the output will be turned into high resistance state.

Power-On Reset---POR, Undervoltage Lockout---UVL

The descriptions in this section assume temperature = 25°C, no output load (RL, CL) , and no significant magnetic field is present.

Power-Up. At power-up, as VCC ramps up, the output is in the following power supply voltage state. When VCC exceeds VPORH, the chip will enter the handshake protocol state. When VCC exceeds VUVLH, the output will go to 1/2*VCC or 2.5V, at this time, the chip is in normal working state.

Power-Down. If VCC drops below VUVLL, the output will be in a high-impedance state. If VCC drops below VPORL, the output is in the following power supply voltage state (See Figure. 11).

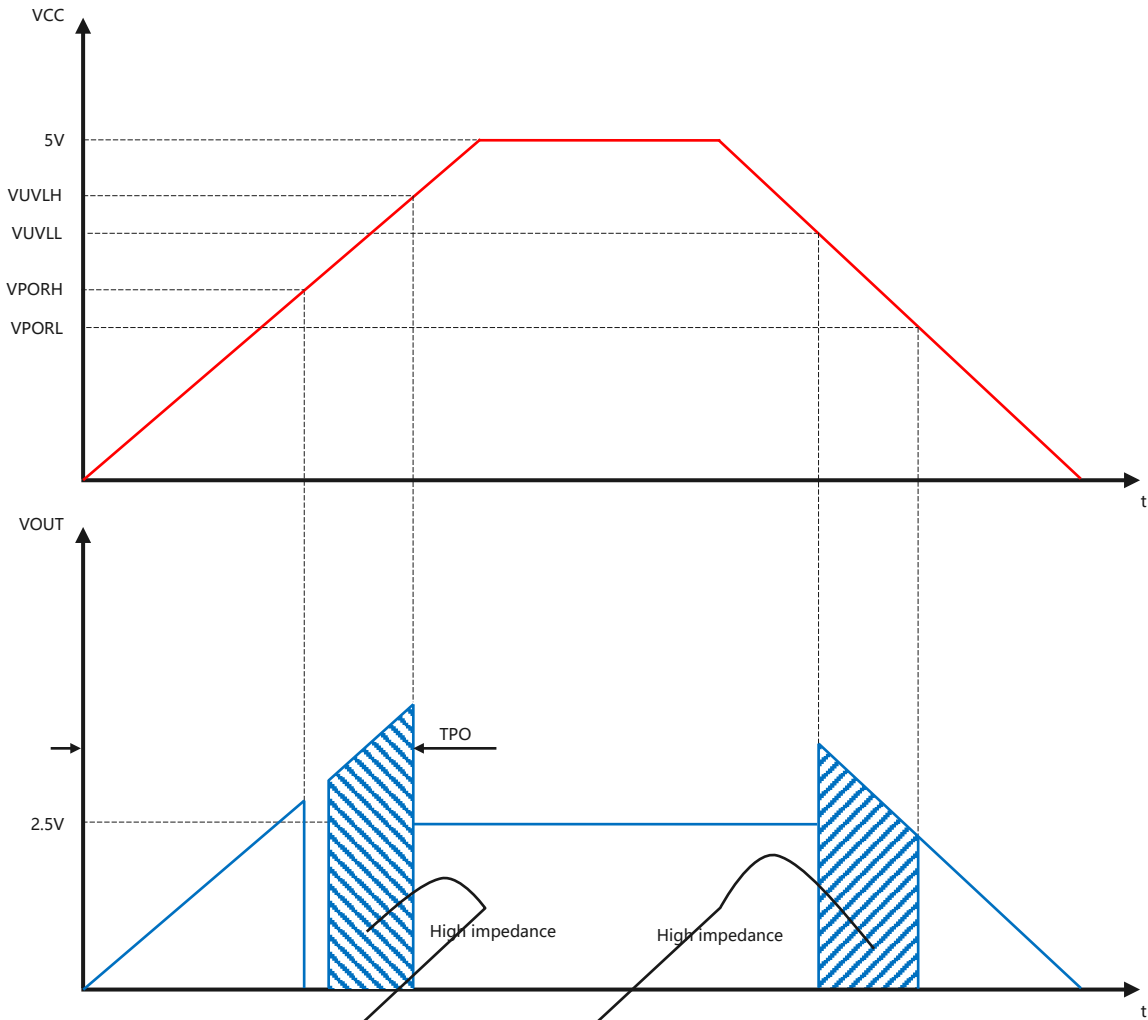
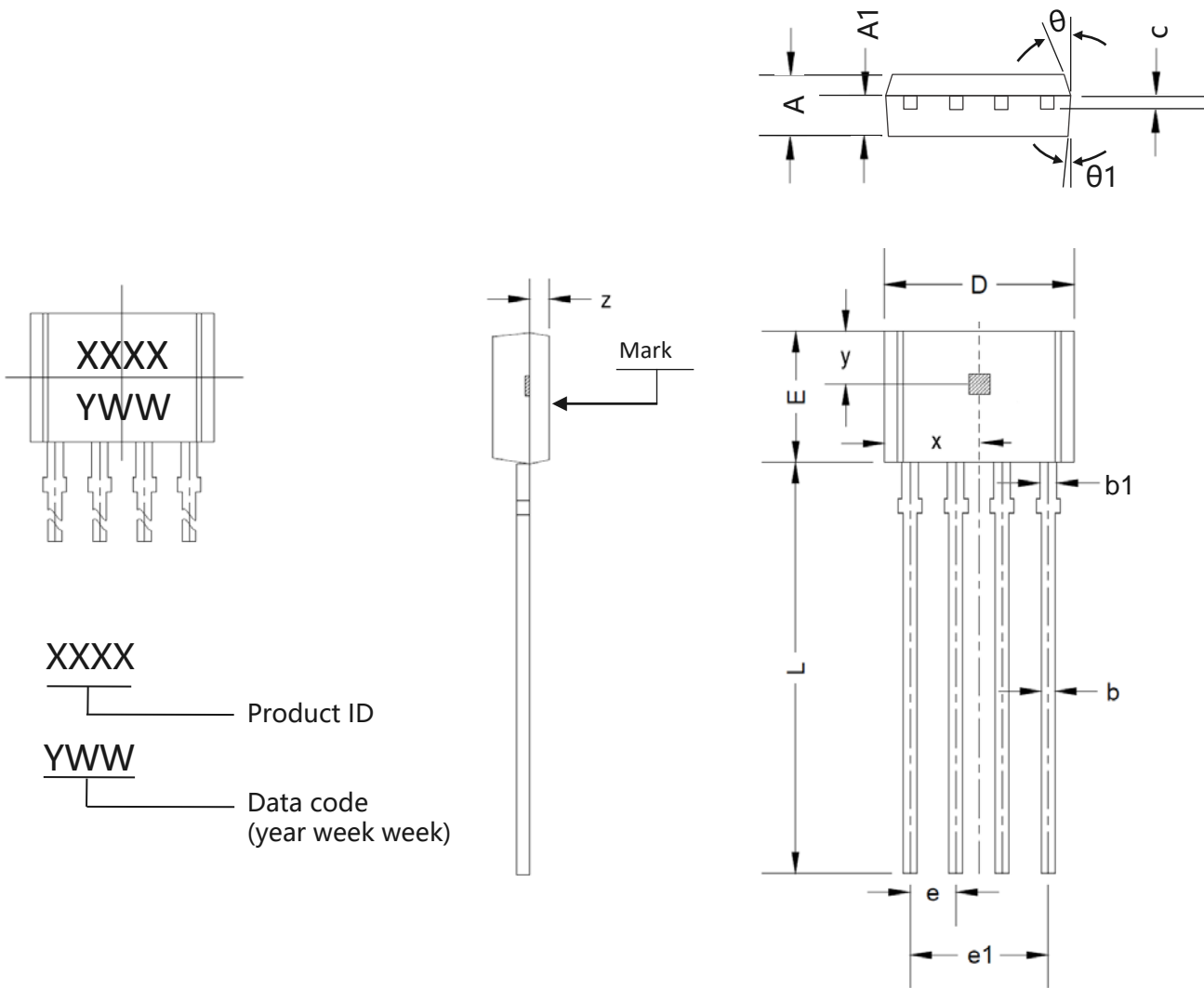


Figure.11 POR and UVL Definition

11 Package Material Information (For Reference Only – Not for Tooling Use)

11.1 SIP-4 Package Information



Symbol	Dimensions in Millimeters		Dimensions in Inches	
	Min	Max	Min	Max
A	1.460	1.660	0.057	0.065
A1	0.660	0.860	0.026	0.034
b	0.350	0.560	0.014	0.022
b1	0.380	0.550	0.015	0.022
c	0.360	0.510	0.014	0.020
D	5.120	5.320	0.202	0.209
E	3.550	3.750	0.140	0.148
e	1.270(BSC)		0.050(BSC)	
e1	3.810(BSC)		0.150(BSC)	
L	13.500	15.500	0.531	0.610
x	2.565(BSC)		0.101(BSC)	
y	0.772(BSC)		0.030(BSC)	
z	0.500(BSC)		0.020(BSC)	
θ	11°		11°	
θ1	6°		6°	

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